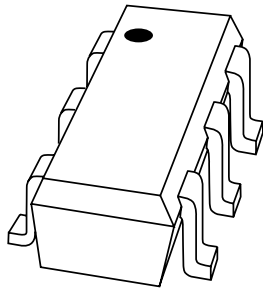


# DATA SHEET



## **BGA2771** MMIC wideband amplifier

Product specification  
Supersedes data of 2001 Oct 19

2002 Aug 06

# MMIC wideband amplifier

# BGA2771

### FEATURES

- Internally matched
- Wide frequency range
- Very flat gain
- High output power
- High linearity
- Unconditionally stable.

### APPLICATIONS

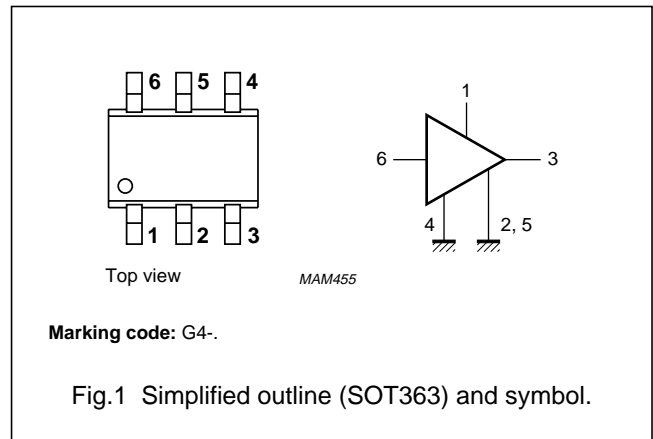
- Cable systems
- LNB IF amplifiers
- General purpose
- ISM.

### DESCRIPTION

Silicon Monolithic Microwave Integrated Circuit (MMIC) wideband amplifier with internal matching circuit in a 6-pin SOT363 SMD plastic package.

### PINNING

PIN	DESCRIPTION
1	V <sub>S</sub>
2, 5	GND2
3	RF out
4	GND1
6	RF in



### QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	TYP.	MAX.	UNIT
V <sub>S</sub>	DC supply voltage		3	4	V
I <sub>S</sub>	DC supply current		33.3	–	mA
S <sub>21</sub>   <sup>2</sup>	insertion power gain	f = 1 GHz	21.4	–	dB
NF	noise figure	f = 1 GHz	4.5	–	dB
P <sub>L(sat)</sub>	saturated load power	f = 1 GHz	13.2	–	dBm

### CAUTION

This product is supplied in anti-static packing to prevent damage caused by electrostatic discharge during transport and handling. For further information, refer to Philips specs.: SNW-EQ-608, SNW-FQ-302A and SNW-FQ-302B.

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**LIMITING VALUES**

In accordance with the Absolute Maximum Rating System (IEC 60134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_S$	DC supply voltage	RF input AC coupled	–	4	V
$I_S$	supply current		–	50	mA
$P_{tot}$	total power dissipation	$T_s \leq 80\text{ °C}$	–	200	mW
$T_{stg}$	storage temperature		–65	+150	°C
$T_j$	operating junction temperature		–	150	°C
$P_D$	maximum drive power		–	10	dBm

**THERMAL CHARACTERISTICS**

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th\ j-s}$	thermal resistance from junction to solder point	$P_{tot} = 200\text{ mW}; T_s \leq 80\text{ °C}$	300	K/W

**CHARACTERISTICS** $V_S = 3\text{ V}; I_S = 33\text{ mA}; f = 1\text{ GHz}; T_j = 25\text{ °C};$  unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_S$	supply current		29	33.3	45	mA
$ S_{21} ^2$	insertion power gain	f = 1 GHz	–	21.4	–	dB
		f = 2 GHz	–	20.8	–	dB
$R_{L\ IN}$	return losses input	f = 1 GHz	–	17	–	dB
		f = 2 GHz	–	13	–	dB
$R_{L\ OUT}$	return losses output	f = 1 GHz	–	9	–	dB
		f = 2 GHz	–	9	–	dB
NF	noise figure	f = 1 GHz	–	4.5	–	dB
		f = 2 GHz	–	4.7	–	dB
BW	bandwidth	at $ S_{21} ^2 -3\text{ dB}$ below flat gain at 1 GHz	–	2.4	–	GHz
$P_{L(sat)}$	saturated load power	f = 1 GHz	–	13.2	–	dBm
		f = 2 GHz	–	10.5	–	dBm
$P_{L\ 1\ dB}$	load power	at 1 dB gain compression; f = 1 GHz	–	12.1	–	dBm
		at 1 dB gain compression; f = 2 GHz	–	8.4	–	dBm
IP3 <sub>(in)</sub>	input intercept point	f = 1 GHz	–	0.5	–	dBm
		f = 2 GHz	–	–4.3	–	dBm
IP3 <sub>(out)</sub>	output intercept point	f = 1 GHz	–	21.9	–	dBm
		f = 2 GHz	–	16.5	–	dBm

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## APPLICATION INFORMATION

Figure 2 shows a typical application circuit for the BGA2771 MMIC. The device is internally matched to 50 Ω, and therefore does not need any external matching. The value of the input and output DC blocking capacitors C2 and C3 should be not more than 100 pF for applications above 100 MHz. However, when the device is operated below 100 MHz, the capacitor value should be increased.

The nominal value of the RF choke L1 is 100 nH. At frequencies below 100 MHz this value should be increased to 220 nH. At frequencies above 1 GHz a much lower value must be used (e.g. 10 nH) to improve return losses. For optimal results, a good quality chip inductor such as the TDK MLG 1608 (0603), or a wire-wound SMD type should be chosen.

Both the RF choke L1 and the 22 nF supply decoupling capacitor C1 should be located as closely as possible to the MMIC.

Separate paths must be used for the ground planes of the ground pins GND1 and GND2, and these paths must be as short as possible. When using vias, use multiple vias per pin in order to limit ground path inductance.

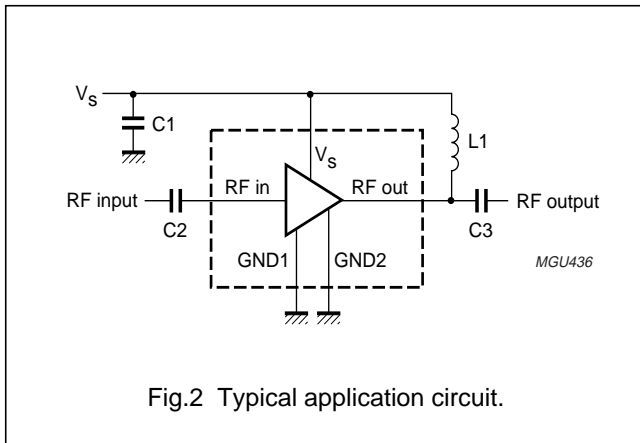


Fig.2 Typical application circuit.

Figure 3 shows two cascaded MMICs. This configuration doubles overall gain while preserving broadband characteristics. Supply decoupling and grounding conditions for each MMIC are the same as those for the circuit of Fig.2.

The excellent wideband characteristics of the MMIC make it an ideal building block in IF amplifier applications such as LBNs (see Fig.4).

As a buffer amplifier between an LNA and a mixer in a receiver circuit, the MMIC offers an easy matching, low noise solution (see Fig.5).

In Fig.6 the MMIC is used as a driver to the power amplifier as part of a transmitter circuit. Good linear performance and matched input and output offer quick design solutions in such applications.

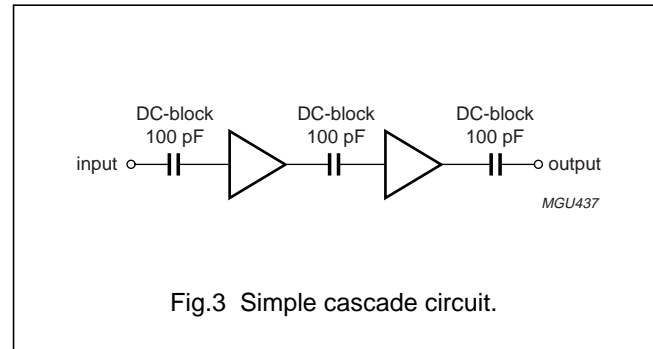


Fig.3 Simple cascade circuit.

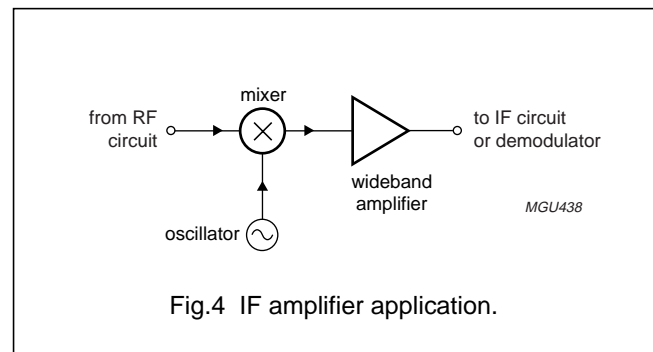


Fig.4 IF amplifier application.

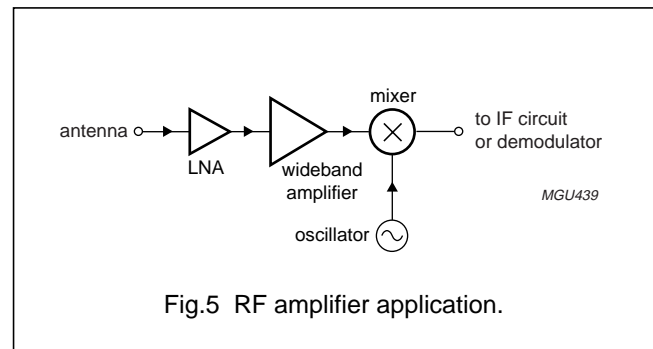


Fig.5 RF amplifier application.

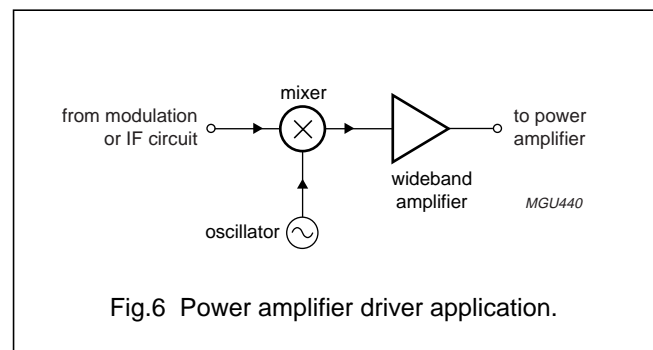
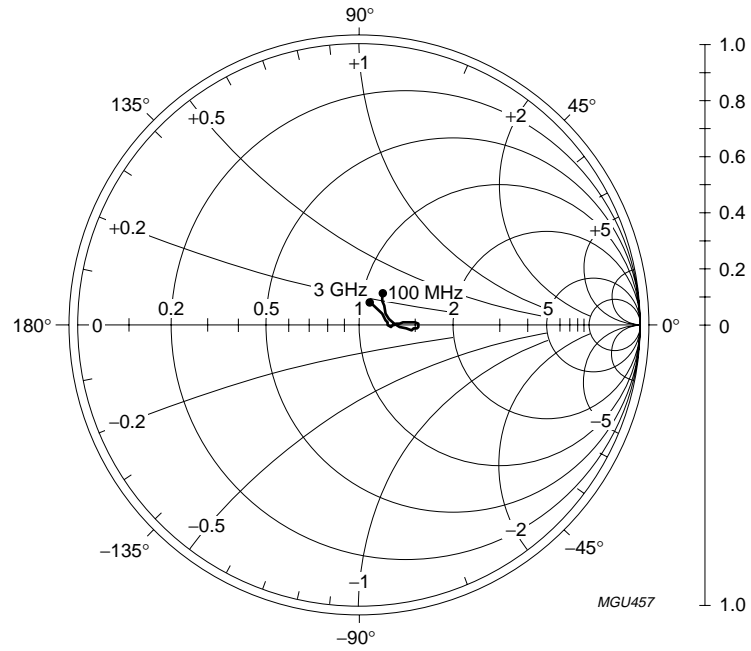


Fig.6 Power amplifier driver application.

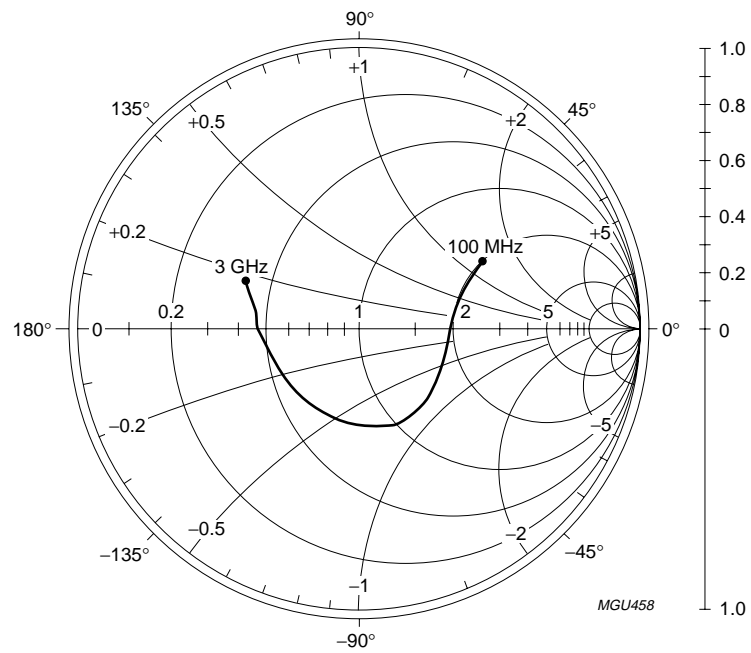
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$I_S = 33.4 \text{ mA}$ ;  $V_S = 3 \text{ V}$ ;  $P_D = -30 \text{ dBm}$ ;  $Z_O = 50 \Omega$ .

Fig.7 Input reflection coefficient ( $s_{11}$ ); typical values.

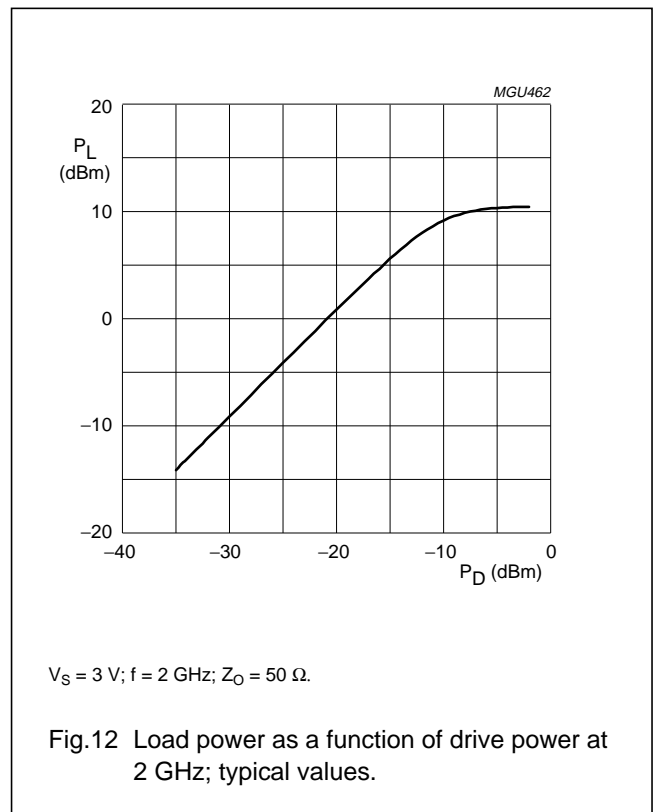
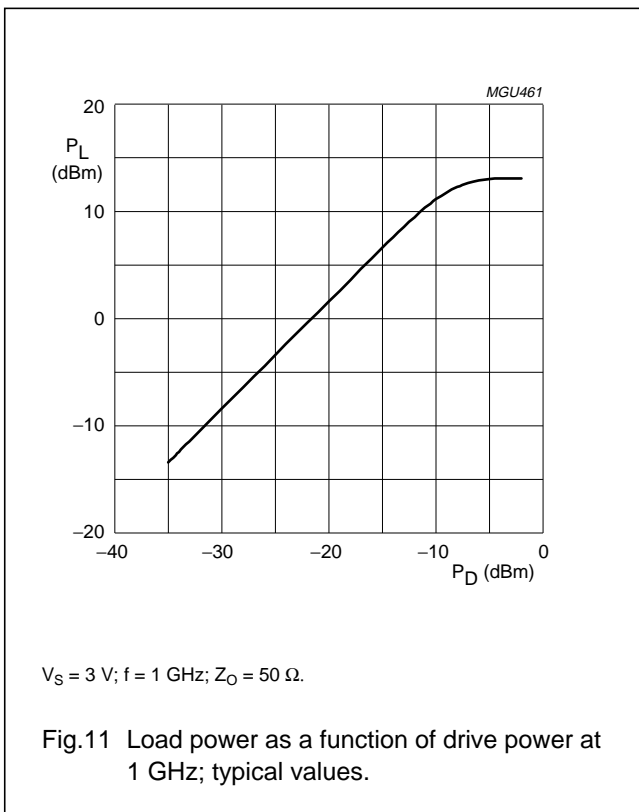
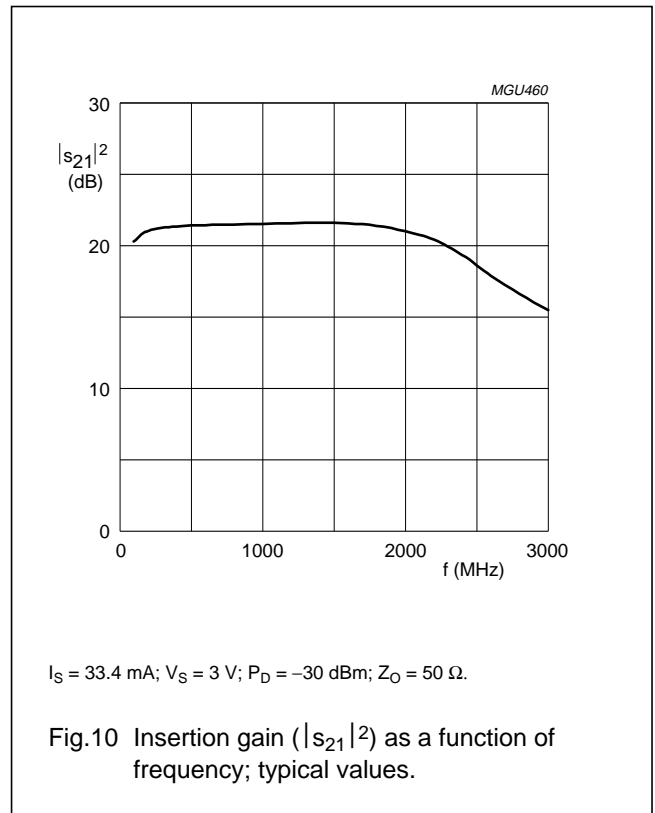
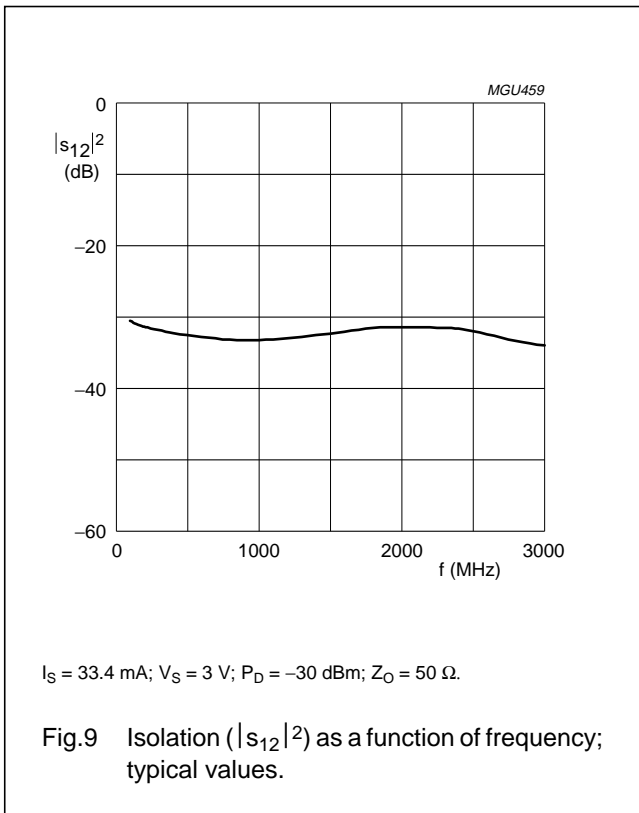


$I_S = 33.4 \text{ mA}$ ;  $V_S = 3 \text{ V}$ ;  $P_D = -30 \text{ dBm}$ ;  $Z_O = 50 \Omega$ .

Fig.8 Output reflection coefficient ( $s_{22}$ ); typical values.

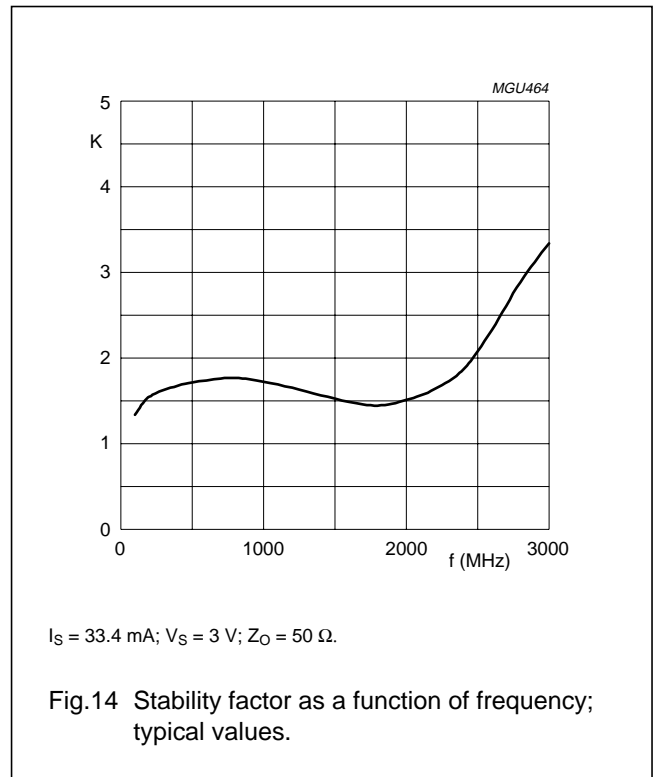
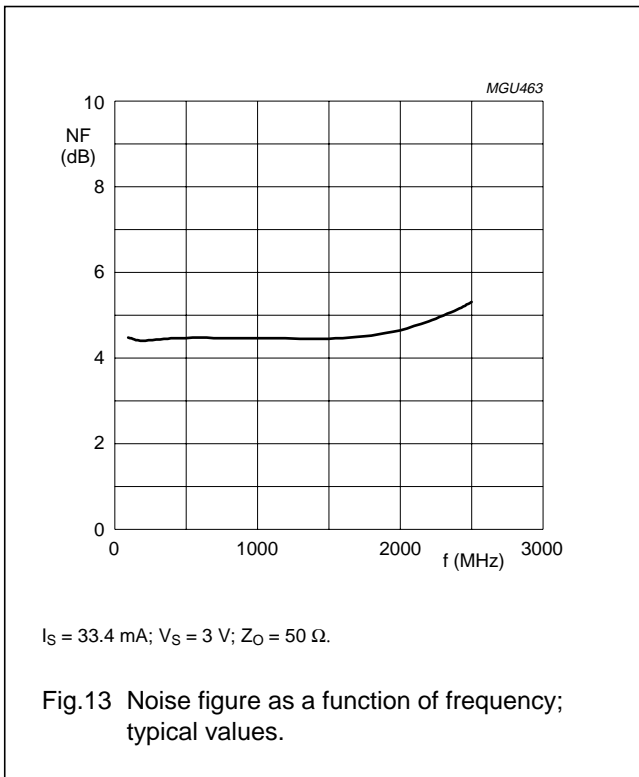
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Scattering parameters

$I_S = 33.4 \text{ mA}; V_S = 3 \text{ V}; P_D = -30 \text{ dBm}; Z_O = 50 \Omega; T_{amb} = 25 \text{ }^\circ\text{C}.$

f (MHz)	S <sub>11</sub>		S <sub>21</sub>		S <sub>12</sub>		S <sub>22</sub>	
	MAGNITUDE (ratio)	ANGLE (deg)	MAGNITUDE (ratio)	ANGLE (deg)	MAGNITUDE (ratio)	ANGLE (deg)	MAGNITUDE (ratio)	ANGLE (deg)
100	0.09328	62.11	10.336	25.98	0.02953	13.71	0.50404	29.78
200	0.09722	25.33	11.266	5.910	0.02687	6.556	0.35904	16.09
400	0.10224	5.155	11.693	-13.69	0.02437	0.870	0.31417	-10.34
600	0.10707	-1.282	11.806	-27.51	0.02288	1.273	0.32541	-29.56
800	0.12009	-0.985	11.851	-40.46	0.02176	3.809	0.34755	-44.52
1000	0.13693	1.692	11.931	-53.00	0.02174	8.643	0.36785	-56.69
1200	0.15676	3.594	11.990	-65.95	0.02229	11.84	0.37169	-68.24
1400	0.17873	4.299	12.036	-79.54	0.02341	13.89	0.36720	-79.76
1600	0.20322	3.166	11.953	-93.52	0.02492	15.56	0.35425	-92.62
1800	0.21560	0.032	11.755	-108.7	0.02645	13.77	0.33802	-107.7
2000	0.20901	-2.617	11.224	-124.7	0.02676	11.10	0.32517	-125.3
2200	0.18846	-5.529	10.499	-140.3	0.02653	9.411	0.32259	-145.1
2400	0.14965	-1.870	9.2991	-156.2	0.02605	6.749	0.33529	-164.4
2600	0.11394	11.81	7.8388	-169.3	0.02388	3.622	0.37019	178.5
2800	0.11394	36.35	6.7932	-178.5	0.02139	6.039	0.39826	165.2
3000	0.13292	50.28	5.9348	174.2	0.01987	12.49	0.44613	156.5

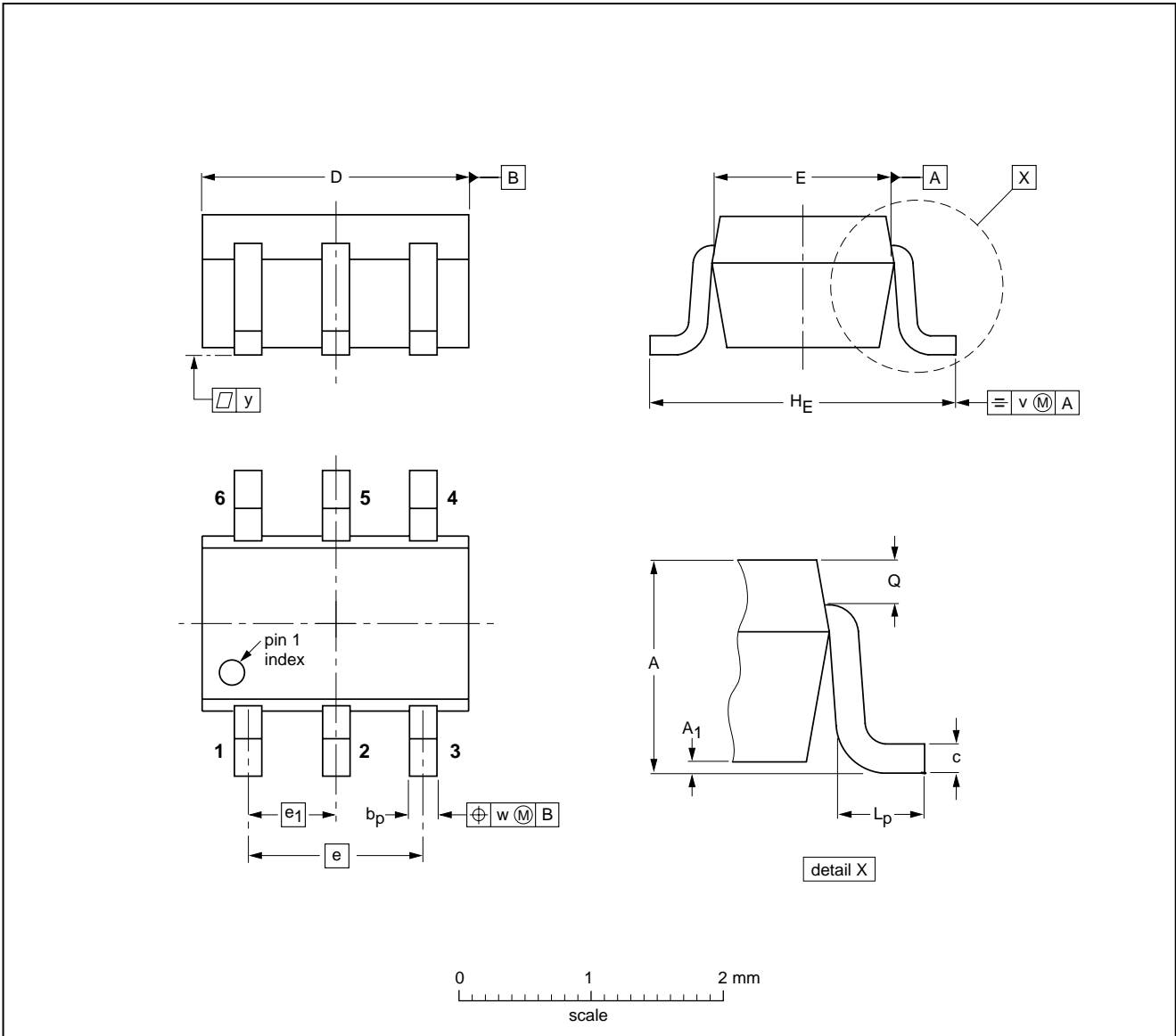
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PACKAGE OUTLINE

Plastic surface mounted package; 6 leads

SOT363



DIMENSIONS (mm are the original dimensions)

UNIT	A	A <sub>1</sub> max	b <sub>p</sub>	c	D	E	e	e <sub>1</sub>	H <sub>E</sub>	L <sub>p</sub>	Q	v	w	y
mm	1.1 0.8	0.1	0.30 0.20	0.25 0.10	2.2 1.8	1.35 1.15	1.3	0.65	2.2 2.0	0.45 0.15	0.25 0.15	0.2	0.2	0.1

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT363			SC-88			97-02-28



## MMIC wideband amplifier

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## DATA SHEET STATUS

DATA SHEET STATUS <sup>(1)</sup>	PRODUCT STATUS <sup>(2)</sup>	DEFINITIONS
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Preliminary data	Qualification	This data sheet contains data from the preliminary specification. Supplementary data will be published at a later date. Philips Semiconductors reserves the right to change the specification without notice, in order to improve the design and supply the best possible product.
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**NOTES**

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**NOTES**

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